Features

- **Universal: Dual Low-Side, Dual High-Side or Half-Bridge Driver**
- **Operating Temperature Range –40 to +125°C**
- **Switching Parameters:**
  - **5-ns Maximum Delay Matching**
  - **8-ns Maximum Pulse-Width Distortion**

  Pai8233B/C/E, Pai8253B/C/E：
  - **19-ns Typical Propagation Delay**
  - **10-ns Minimum Pulse Width**

  Pai8233A/D/F, Pai8253A/D/F：
  - **90-ns Typical Propagation Delay**
  - **50-ns Minimum Pulse Width**
- **Common-Mode Transient Immunity (CMTI)**

  Pai8233B/C/E, Pai8253B/C/E：100kV/us

  Pai8233A/D/F, Pai8253A/D/F：300kV/us
- **Isolation Barrier Life >40 Years**
- **4-A Peak Source, 8-A Peak Sink Output**
- **TTL and CMOS Compatible Inputs**
- **3V to 5.5V Input VCCI Range to Interface with Both Digital and Analog Controllers**
- **Up to 25V VDD Output Drive Supply**
- **Programmable Overlap and Dead Time**
- **Fast Disable for Power Sequencing**
- **Safety-Related Certifications: (Pending)**
  - **3535V_{PK} Basic Isolation per DIN EN IEC 60747-17（VDE 0884-17）:2021-10**
  - **2.5kV_{RMS} Isolation for 1 Minute per UL 1577**
  - **CQC Certification per GB4943.1-2022**
- **RoHS-compliant, LGA13 package**

Applications

- **HEV and BEV Battery Chargers**
- **Isolated Converters in DC-DC and AC-DC Power Supplies**
- **Server, Telecom, IT and Industrial Infrastructures**
- **Motor Drive and DC-to-AC Solar Inverters**
- **LED Lighting**
- **Inductive Heating**
- **Uninterruptible Power Supply (UPS)**

General Description

The Pai8233X/Pai8253X are the dual-channel isolated gate drivers based on *iDivider®* technology of 2Pai Semi. It has a source peak current of 4A and a sink peak current of 8A. The maximum switching frequency can reach 5MHz. It is suitable for gate drive of MOSFET, IGBT and SiC MOSFET.

The input side is isolated from the two output sides by an isolation barrier that can withstand 2.5kV_{RMS} isolation voltage, and the typical common-mode transient immunity (CMTI) capability is 300kV/us. The internal functional isolation between the two secondary side drivers allows a maximum operating voltage of 700V_{DC}.

Every driver can be configured as two low-side drivers two high-side drivers, or a half-bridge driver with programmable dead time (DT). A disable pin shuts down both outputs simultaneously when it is set high and allows normal operation when left open or grounded. As a fail-safe measure, primary-side logic failures force both outputs low.

Each device accepts VDD supply voltages up to 25V. A wide input VCCI range from 3V to 5.5V makes the driver suitable for interfacing with both analog and digital controllers. All the supply voltage pins have under voltage lock-out (UVLO) protection.

Functional Block Diagram

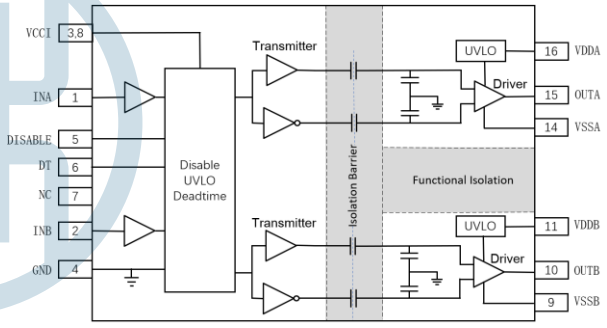
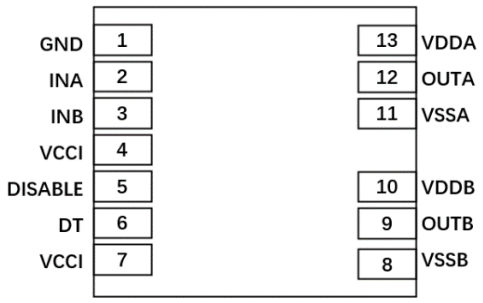


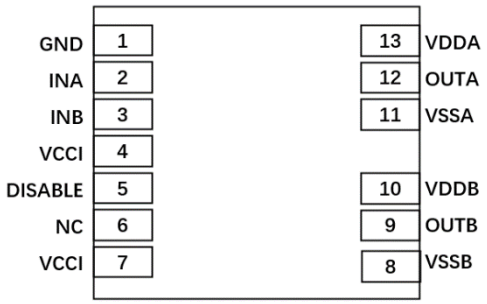
Fig 1. Functional Block Diagram

Room 307-309, No.22, Boxia Road, Pudong New District, Shanghai, 201203, China
021-50850681

1. Pin Configurations and Functions



Pai8233 LGA13



Pai8253 LGA13

Fig 2. Pin Configuration

Table 1. Pin Function Descriptions

PIN NAME	PIN NO.		DESCRIPTION
	Pai8233 LGA13	Pai8253 LGA13	
INA	2	2	Input signal for A channel. INA input has a TTL/CMOS compatible input threshold. This pin is pulled low internally if left open. It is recommended to tie this pin to ground if not used to achieve better noise immunity.
INB	3	3	Input signal for B channel. INB input has a TTL/CMOS compatible input threshold. This pin is pulled low internally if left open. It is recommended to tie this pin to ground if not used to achieve better noise immunity.
VCCI	4,7	4,7	Primary-side supply voltage. Locally decoupled to GND using a low ESR/ESL capacitor located as close to the device as possible.
GND	1	1	Primary-side ground reference. All signals in the primary side are referenced to this ground.
DISABLE	5	5	Disables both driver outputs if asserted high, enables if set low or left open. This pin is pulled low internally if left open. If this pin is not used, it is recommended to ground this pin to obtain better noise immunity. When connecting to a microcontroller, use a low ESR/ESL capacitor of approximately 1nF to bypass the DIS pin.
DT	6	/	Programmable dead time function. Tying DT to VCCI allows the outputs to overlap. Leaving DT open sets, the dead time to <15 ns. Placing a 500-Ω to 500-kΩ resistor (RDT) between DT and GND adjusts dead time according to: DT (in ns) ≈ 10 x R_{DT} (in kΩ).
NC	/	6	No Internal connection.
VSSB	8	8	Ground for secondary-side driver B. Ground reference for secondary side B channel.
OUTB	9	9	Output of driver B. Connect to the gate of the B channel FET or IGBT.
VDDB	10	10	Secondary-side power for driver B. Locally decoupled to VSSB using low ESR/ESL capacitor located as close to the device as possible.
VSSA	11	11	Ground for secondary-side driver A. Ground reference for secondary side A channel.
OUTA	12	12	Output of driver A. Connect to the gate of the A channel FET or IGBT.
VDDA	13	13	Secondary-side power for driver A. Locally decoupled to VSSA using a low ESR/ESL capacitor located as close to the device as possible.

2. Specifications

2.1.Absolute Maximum Ratings

Table 2.Absolute Maximum Ratings (1)

PARAMETER	SYMBOL	MIN	MAX	UNIT
Input power supply voltage	VCCI to GND	–0.6	7	V
Driver bias supply	VDDA-VSSA, VDDB-VSSB	–0.4	30	V
Input signal voltage	INA, INB, DIS, DT to GND	–0.6	VCCI+0.5	V
Input signal voltage	INA, INB Transient for 50ns	–5	VCCI+0.5	V
Output signal voltage	OUTA to VSSA, OUTB to VSSB	–0.4	VDDA+0.5, VDDB+0.5	V
Output signal voltage	OUTA to VSSA, OUTB to VSSB, Transient for 200 ns	–2	VDDA+0.5, VDDB+0.5	V
Channel to channel voltage	VSSA-VSSB, VSSB-VSSA LGA13 Package	–700	700	V
Junction temperature, T_J (2)	T_J	–40	150	℃
Storage temperature, T_{stg}	T_{stg}	–65	150	℃

1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
2) To maintain the recommended operating conditions for T_J, see the Thermal Information.

2.2.ESD Caution

ESD(Electrostatic Discharge) Sensitive device
Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

2.3.Recommended Operating Conditions

Table 3.Over operating free-air temperature range (unless otherwise noted)

SYMBOL	DESCRIPTION	MIN	MAX	UNIT
VCCI	VCCI Input supply voltage	3	5.5	V
VDDA-VSSA, VDDB-VSSB	Driver output bias supply---6V Version	6.5	25	V
	Driver output bias supply---9V Version	9.2	25	V
	Driver output bias supply---12V Version	13.5	25	V
VIA-GND, VIB-GND, DISABLE-GND,	input voltage	0	V_{CCI}	V
VOA-VSSA, VOB-VSSB	output voltage	0	VDDA/VDDB	V
VSSA-VSSB/VSSA-GND/VSSB-GND	Channel to channel working voltage	-700	700	V
t_{INA}, t_{INB}	B/C/E Valid pulse width	10	/	ns
	A/D/F Valid pulse width	50	/	ns
T_A	Ambient Temperature under Bias	-40	125	℃
T_J	Junction Temperature	–40	130	℃

2.4.Thermal Information

Table 4.Thermal Information

SYMBOL	PARAMETER	TYP	Unit
$R_{\theta JA}$	Junction-to-ambient thermal resistance	207	°C/W

2.5.Power Ratings

Table 5.Power Ratings

VCCI = 3.3 V, VDDA/B = 12 V, INA/B = 3.3 V, 2MHz 50% duty cycle square wave 1nF load

SYMBOL	PARAMETER	VALUE	UNIT
PD	Power dissipation	0.603	W
PDI	Power dissipation by transmitter side	0.003	W
PDA, PDB	Power dissipation by each driver side	0.30	W

3. Insulation Specifications

Table 6.Insulation Specifications

PARAMETER	DESCRIPTION	TEST CONDITIONS	VALUE	UNIT
CLR	External clearance	Shortest pin-to-pin distance through air	>3.5	mm
CPG	External creepage	Shortest pin-to-pin distance across the package surface	>3.5	mm
DTI	Distance through insulation	Minimum internal gap (internal clearance) of the double insulation (2 × 10.5 µm)	>21	µm
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	> 400	V
Material group	Material group	According to IEC 60664-1	II	
Overvoltage category per IEC 60664-1		Rated mains voltage ≤150 VRMS	I-III	
		Rated mains voltage ≤ 300 VRMS	I-II	
		Rated mains voltage ≤ 600 VRMS	I	
		Rated mains voltage ≤ 1000 VRMS	/	

Table 7. DIN EN IEC 60747-17 （VDE 0884-17）:2021-10

PARAMETER	DESCRIPTION	TEST CONDITIONS	VALUE	UNIT
V_{IORM}	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	792	V_{PK}
V_{IOWM}	Maximum working isolation voltage	AC voltage (sine wave)	560	V_{RMS}
		DC voltage	792	V_{DC}
V_{IOTM}	Maximum transient isolation voltage	$V_{TEST} = V_{IOTM}$, t = 60 sec (qualification) $V_{TEST} = 1.2 \times V_{IOTM}$, t = 1 s (100% production)	3535	V_{PK}
V_{IOSM}	Maximum surge isolation voltage	Test method per IEC 62368-1, 1.2/50 µs waveform, $V_{TEST} = 1.3 \times V_{IOSM}$ (qualification)	3500	V_{PK}
q_{pd}	Apparent charge	Method a, After Input/Output safety test subgroup 2/3. $V_{ini} = V_{IOTM}$, t_{ini} = 60s;$V_{pd(m)} = 1.2 \times V_{IORM}$, t_m = 10s	<5	pC
		Method a, After environmental tests subgroup 1.V_{ini} = V_{IOTM}, t_{ini} = 60s;$V_{pd(m)} = 1.3 \times V_{IORM}$, t_m = 10s	<5	pC
		Method b1; At routine test (100% production) and preconditioning (type test) $V_{ini} = 1.2 \times V_{IOTM}$; t_{ini} = 1s;$V_{pd(m)} = 1.5 * V_{IORM}$, t_m = 1s	<5	pC

PARAMETER	DESCRIPTION	TEST CONDITIONS	VALUE	UNIT
C_{IO}	Barrier capacitance, input to output	V_{IO} = 0.4 sin (2πft), f =1 MHz	1.2	pF
R_{IO}	Isolation resistance, input to output	V_{IO} = 500 V at T_A = 25°C	> 10^{12}	> 10^{12}
		V_{IO} = 500 V at 100°C ≤ T_A ≤ 125°C	> 10^{11}	> 10^{11}
		V_{IO} = 500 V at T_S =150°C	> 10^{9}	> 10^{9}
Pollution degree			2	
Climatic category			40/125/21	

Table 8.UL 1577

PARAMETER	DESCRIPTION	TEST CONDITIONS	VALUE	UNIT
V_{ISO}	Withstand isolation voltage	V_{TEST} = V_{ISO}, t = 60 sec.(qualification), V_{TEST} = 1.2 × V_{ISO}, t = 1 sec (100% production)	2500	V_{RMS}

3.1.Safety-Related Certifications

Table 9.Safety-Related Certifications (Pending)

CQC	Certified according to GB 4943.1-2022	Reinforced Insulation, Altitude ≤ 5000 m, Tropical Climate 660 V_{RMS} maximum working voltage.	Certificate Number: Pending
UL	Recognized under UL 1577 5th Ed Component Recognition Program CSA Component Acceptance Service Notice No. 5A	Single protection, 2500V_{RMS}	Certificate Number: Pending
VDE	Certified according to DIN V VDE V 0884-11:2017-01, and DIN EN 60950-1 (VDE 0805 Teil 1):2014-08	Basic Insulation Maximum Transient Isolation voltage, 3535V_{PK}; Maximum Repetitive Peak Isolation Voltage,792V_{PK}; Maximum Surge Isolation Voltage,3500 V_{PK}	Certification number: Pending

3.2.Safety-Limiting Values

Safety limiting intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

Table 10.Safety-Limiting Values

Symbol	PARAMETER	TEST CONDITIONS	SIDE	MAX	UNIT
I_S	Safety output supply current	$R_{\theta JA}$ = 207°C/W, VDDA/B = 12 V, T_A = 25°C, T_J = 150°C	Driver A Driver B	25	mA
		$R_{\theta JA}$ =207°C/W, VDDA/B = 25 V, T_A = 25°C, T_J = 150°C	Driver A Driver B	12	mA
P_S	Safety supply power	$R_{\theta JA}$ = 207℃/W, T_A = 25℃, T_J = 150℃	TOTAL	600	mW
T_S	Safety temperature[1]			150	°C

1) The maximum safety temperature, T_S, has the same value as the maximum junction temperature, T_J, specified for the device. The I_S and P_S parameters represent the safety current and safety power respectively. The maximum limits of I_S and P_S should not be exceeded. These limits vary with the ambient temperature, T_A.
The junction-to-air thermal resistance, $R_{\theta JA}$, in the Thermal Information table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:
$T_J = T_A + R_{\theta JA} \times P$, where P is the power dissipated in the device.
$T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$, where $T_{J(max)}$ is the maximum allowed junction temperature.$P_S = I_S \times V_I$, where V_I is the maximum input voltage.

4. Truth table

Table 11.Truth table

Input		DISABLE	VCCI/VDDA/VDDB Condition			Output		Notes
INA	INB		VCCI	VDDA	VDDB	VOA	VOB	
L	L	L or Left Open	Powered	Powered	Powered	L	L	If Dead Time function is used, output transitions occur after the dead time expires. See Programmable Dead Time (DT) Pin[2]
L	H	L or Left Open	Powered	Powered / Unpowered	Powered	L	H[2]	
H	L	L or Left Open	Powered	Powered	Powered / Unpowered	H[2]	L	
H	H	L or Left Open	Powered	Powered	Powered	L	L	DT is left open or programmed with R_{DT}
H	H	L or Left Open	Powered	Powered	Powered	H[2]	H[2]	DT pin pulled to VCCI.
X[1]	X[1]	H	Powered	Powered	Powered	L	L	
L	L	L or Left Open	Unpowered	Powered	Powered	L	L	

Notes：

1) "X" means L, H or left open.
2) The VOA/VOB output is high only in this case. If there are other conditions, VOA / VOB output is low.
$T_{DT}\approx 10^{*}\ R_{DT}$ (T_{DT} is in nS and R_{DT} is in kΩ).Connect DT to VCCI to overlap the output. Keep DT open to set the dead time to <15ns.Place a 500Ω to 500kΩ resistor (R_{DT}) between DT and GND to adjust the dead time according to the following conditions: DT(ns) ≈10 x R_{DT}(kΩ).

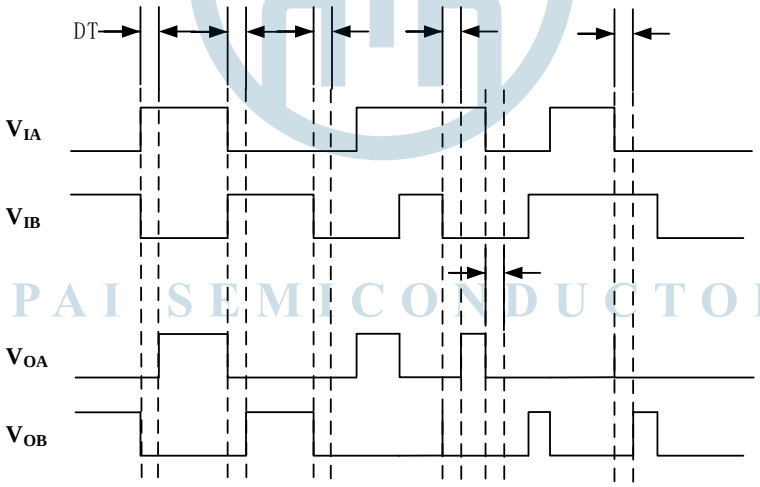


Fig 3.Input/Output Timing Diagram

5. Specifications

5.1.Electrical Characteristics

Table 12.ELECTRICAL CHARACTERISTICS

VCCI = 3.3 V or 5 V, 0.1-µF capacitor from VCCI to GND, Pai823(5)3A/B/C/D VDDA = VDDB = 12 V, Pai823(5)3E/F VDDA = VDDB = 15V, 1-µF capacitor from VDDA and VDDB to VSSA and VSSB, T_A = −40 to +125℃, (unless otherwise noted)

Parameter	Description	Test Conditions	Min	Typ	Max	Unit
I_{VCCI}	VCCI quiescent current	V_{INA} = 0 V, V_{INB} = 0 V		0.5	1.1	mA
I_{VDDA}, I_{VDDB}	VDDA and VDDB quiescent current	V_{INA} = 0 V, V_{INB} = 0 V		0.9	1.5	mA
I_{VCCI}	VCCI operating current	VINA, VINB input signal f = 500 kHz, 50% duty cycle, each channel load capacitance 100 pF ± 20%, including the input capacitance of the measuring instrument.		0.9		mA
I_{VDDA}, I_{VDDB}	VDDA and VDDB operating current			2.2		mA
V_{VCCI_ON}	VCCI UVLO Rising threshold		2.55	2.7	2.85	V
V_{VCCI_OFF}	VCCI UVLO Falling threshold		2.35	2.5	2.65	V
V_{VCCI_HYS}	UVLO Threshold hysteresis			0.2		V
V_{VDDA_ON}, V_{VDDB_ON}	UVLO Rising threshold---9V Version (Pai8233A/C, Pai8253A/C)		8.3	8.7	9.2	V
V_{VDDA_OFF}, V_{VDDB_OFF}	UVLO Falling threshold---9V Version (Pai8233A/C, Pai8253A/C)		7.8	8.2	8.7	V
V_{VDDA_HYS} V_{VDDB_HYS}	UVLO Threshold hysteresis---9V Version (Pai8233A/C, Pai8253A/C)			0.5		V
V_{VDDA_ON}, V_{VDDB_ON}	UVLO Rising threshold---6V Version (Pai8233B/D, Pai8253B/D)		5.7	6.1	6.5	V
V_{VDDA_OFF}, V_{VDDB_OFF}	UVLO Falling threshold---6V Version (Pai8233B/D, Pai8253B/D)		5.4	5.8	6.2	V
V_{VDDA_HYS} V_{VDDB_HYS}	UVLO Threshold hysteresis---6V Version (Pai8233B/D, Pai8253B/D)			0.3		V
V_{VDDA_ON}, V_{VDDB_ON}	UVLO Rising threshold---12V Version (Pai8233E/F, Pai8253E/F)		11.9	12.8	13.7	V
V_{VDDA_OFF}, V_{VDDB_OFF}	UVLO Falling threshold---12V Version (Pai8233E/F, Pai8253E/F)		10.9	11.8	12.7	V
V_{VDDA_HYS} V_{VDDB_HYS}	UVLO Threshold hysteresis---12V Version (Pai8233E/F, Pai8253E/F)			1.0		V
V_{INAH}, V_{INBH}, V_{DISH}	Input high threshold voltage		1.6	1.8	2	V
V_{INAL}, V_{INBL}, V_{DISL}	Input low threshold voltage		0.8	1	1.2	V
V_{INA_HYS}, V_{INB_HYS}, V_{DIS_HYS}	Input threshold hysteresis			0.8		V
V_{INA}, V_{INB}	Negative transient, ref to GND, 50 ns pulse	Not production tested, bench test only	–5			V
I_{OA+}, I_{OB+}	Peak output source current	C_{VDD} = 10 µF, C_{LOAD} = 0.68 µF, f= 100Hz, bench measurement		4		A
I_{OA-}, I_{OB-}	Peak output sink current			8		A
R_{OHA}, R_{OHB}	Output resistance at high state	IOUT = –10 mA, TA = 25°C, R_{OHA}, R_{OHB} do not represent drive pull- up performance. See t_{RISE} in Switching Characteristics and Output Stage for details.		1		Ω
R_{OLA}, R_{OLB}	Output resistance at low state	IOUT = 10 mA, TA = 25°C		0.4		Ω
V_{OHA}, V_{OHB}	Output voltage at high state	VDDA, VDDB = 15 V, IOUT = –10 mA, TA = 25°C		14.99		V
V_{OLA}, V_{OLB}	Output voltage at low state	VDDA, VDDB = 15 V, IOUT = 10 mA, TA = 25°C		4		mV
DT	Dead time (Pai8233X)	Pull DT pin to VCCI	Overlap determined by INA INB			ns
		DT pin is left open, min spec characterized only, tested for outliers	0	8	15	ns
		R_{DT} = 20 kΩ	150	200	250	ns
T_{JF}	Junction Temperature Shutdown, Falling Edge			140		°C
T_{JR}	Junction Temperature Shutdown, Rising Edge			150		°C

5.2. Switching Characteristics

Table 13.Switching Characteristics

VCCI = 3.3 V or 5 V, 0.1-µF capacitor from VCCI to GND, Pai823(5)3A/B/C/D VDDA = VDDB = 12 V, Pai823(5)3E/F VDDA = VDDB = 15V, 1-µF capacitor from VDDA and VDDB to VSSA and VSSB, no load, TA = –40°C to +125°C, (unless otherwise noted).

Parameter	Description	Test Conditions	Min	Typ	Max	Unit
t_r	Output rise time	C_{OUT} = 1.8 nF		7	18	ns
t_f	Output fall time	C_{OUT} = 1.8 nF		8	12	ns
t_{PWmin}	Minimum pulse width	Output off for less than minimum, C_{OUT} = 0 pF		10	20	ns
t_{PDHL}	Pai823(5)3B/C/E Propagation delay from INx to OUTx falling edges		14	19	30	ns
t_{PDLH}	Pai823(5)3B/C/E Propagation delay from INx to OUTx rising edges		14	19	30	ns
t_{PWD}	Pai823(5)3B/C/E Pulse width distortion $\lvert t_{PDLH} - t_{PDHL} \rvert$				5	ns
t_{PDHL}	Pai823(5)3A/D/F Propagation delay from INx to OUTx falling edges		60	90	150	ns
t_{PDLH}	Pai823(5)3A/D/F Propagation delay from INx to OUTx rising edges		60	90	150	ns
t_{PWD}	Pai823(5)3A/D/F Pulse width distortion $\lvert t_{PDLH} - t_{PDHL} \rvert$				8	ns
t_{DM}	Propagation delays matching between VOUTA, VOUTB	f = 100 kHz			5	ns
$\lvert CM_H \rvert$	Pai823(5)3B/C/E High-level common-mode transient immunity	INA and INB both are tied to VCCI; V_{CM}=1000V;		100		V/ns
$\lvert CM_L \rvert$	Pai823(5)3B/C/E Low-level common-mode transient immunity	INA and INB both are tied to GND; V_{CM}=1000V;		100		V/ns
$\lvert CM_H \rvert$	Pai823(5)3A/D/F High-level common-mode transient immunity	INA and INB both are tied to VCCI; V_{CM}=1500V;	300			V/ns
$\lvert CM_L \rvert$	Pai823(5)3A/D/F Low-level common-mode transient immunity	INA and INB both are tied to GND; V_{CM}=1500V;	300			V/ns
t_{SD}	Pai823(5)3B/C/E Shutdown Time from Disable True				40	ns
$t_{RESTART}$	Pai823(5)3B/C/E Restart Time from Disable False				40	ns
t_{SD}	Pai823(5)3A/D/F Shutdown Time from Disable True				150	ns
$t_{RESTART}$	Pai823(5)3A/D/F Restart Time from Disable False				150	ns
t_{START}	Device Start-up Time	Time from VDD_ = VDD_UV+ to VOA, VOB = VIA, VIB			100	µs

5.3.Insulation Characteristics Curves

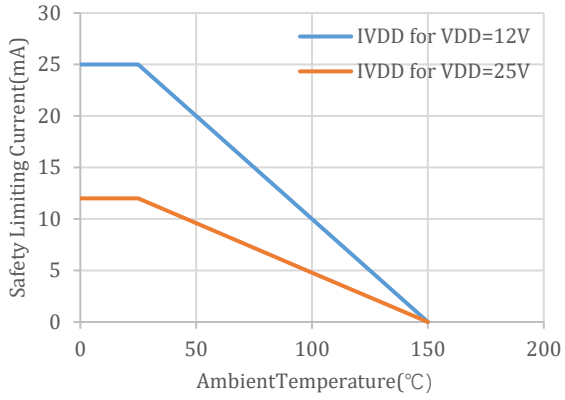


Fig 4.Thermal Derating Curve for Limiting Current Per VDE

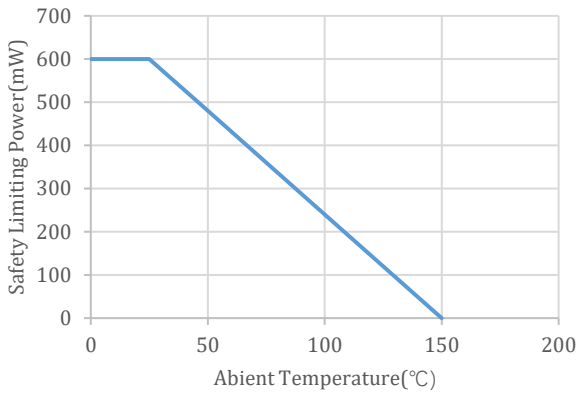


Fig 5.Thermal Derating Curve for Limiting Power Per VDE

5.4.Typical Characteristics

VDDA = VDDB= 15 V, VCCI = 3.3 V, T_A = 25°C, No load, unless otherwise noted.

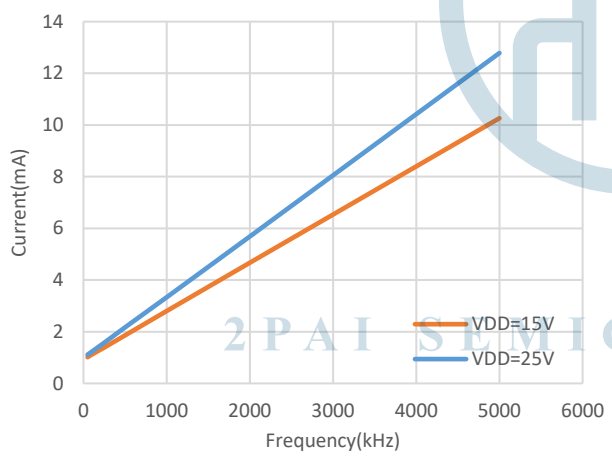


Fig 6.Per Channel Current ($I_{VDDA/B}$) (No load)

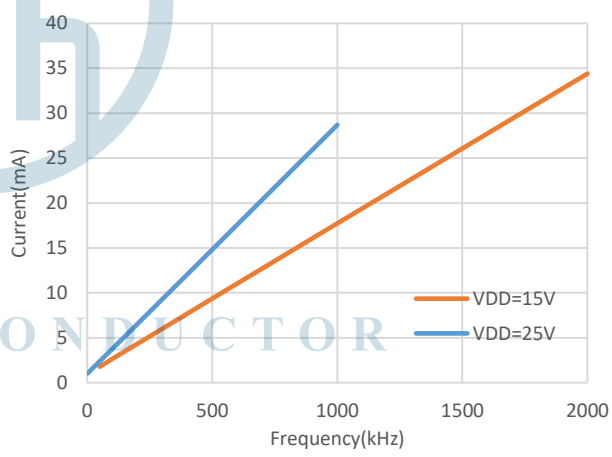


Fig 7.Per Channel Current ($I_{VDDA/B}$) (1nF Load)

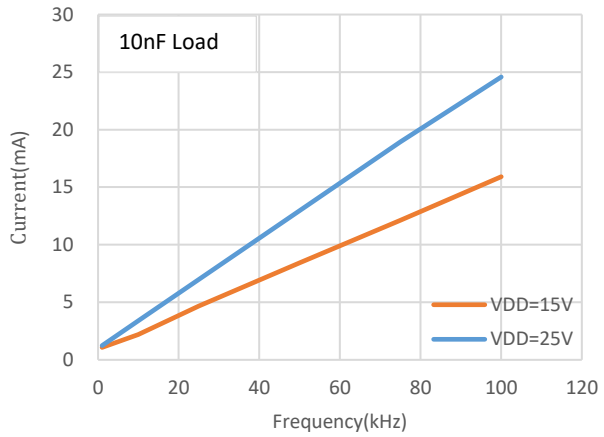


Fig 8.Per Channel Current ($I_{VDDA/B}$) (10nF Load)

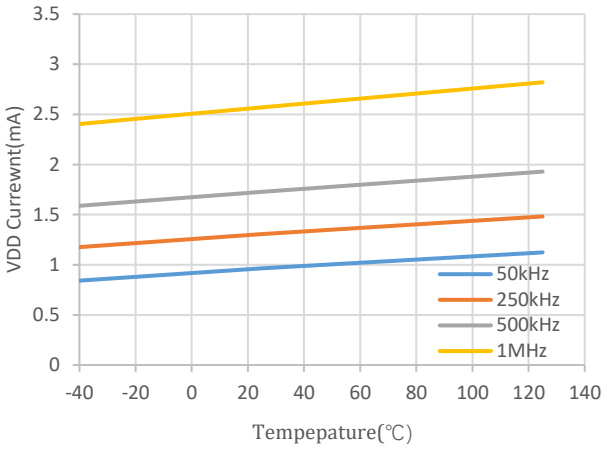


Fig 9.Per Channel Current ($I_{VDDA/B}$) (No Load)

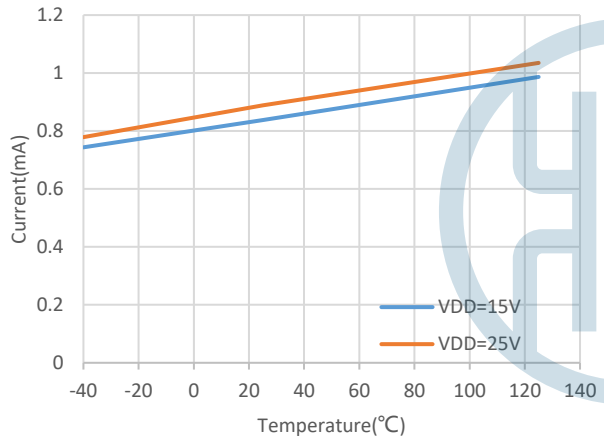


Fig 10.Per Channel Quiescent Current ($I_{VDDA/B}$)

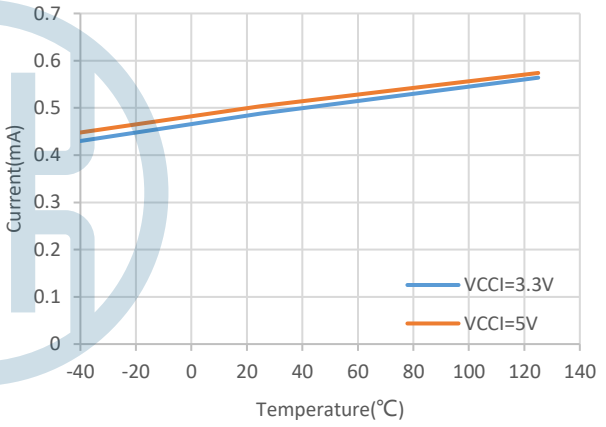


Fig 11.I_{VCCI} Quiescent Current

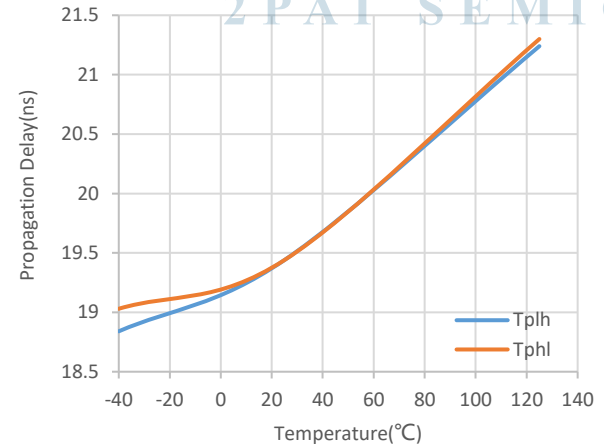


Fig 12.Pai823(5)3B/C/E Propagation Delay

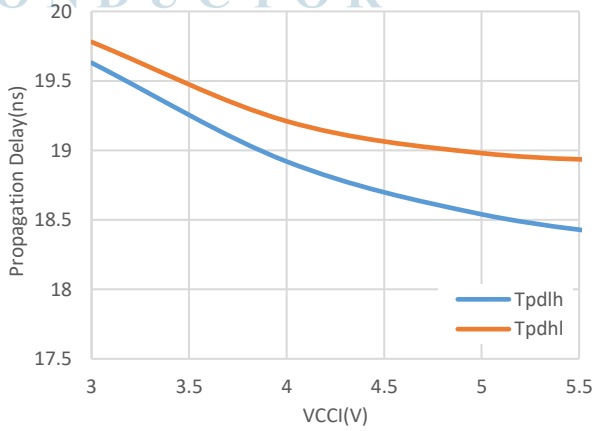


Fig 13.Pai823(5)3B/C/E Propagation Delay

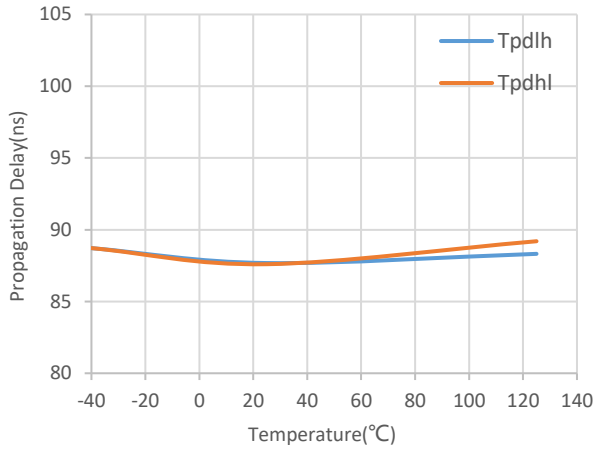


Fig 14.Pai823(5)3A/D/F Propagation Delay

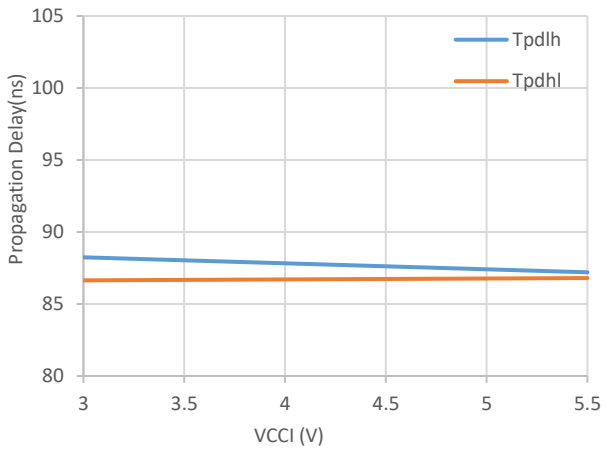


Fig 15. Pai823(5)3A/D/F Propagation Delay

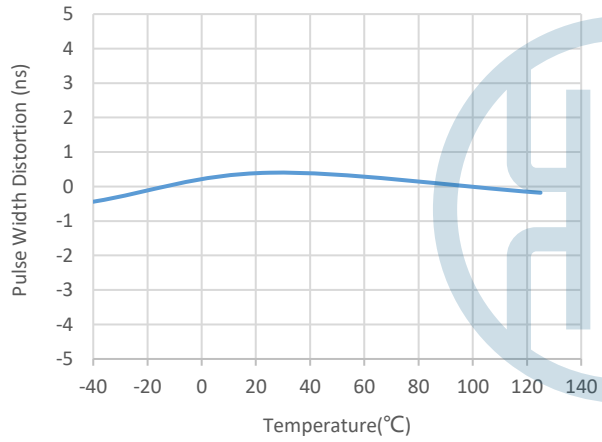


Fig 16.Pulse Width Distortion

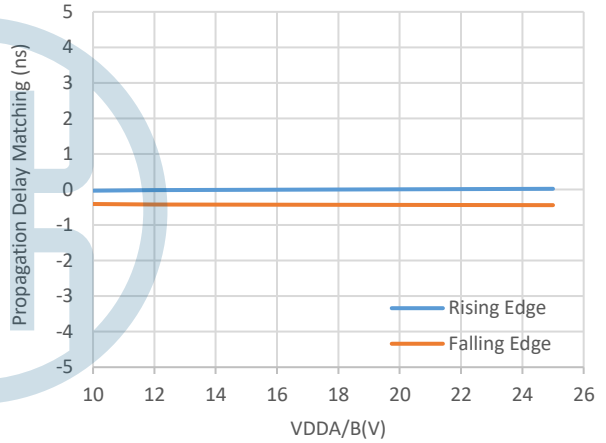


Fig 17.Propagation Delay Matching (t_{DM})

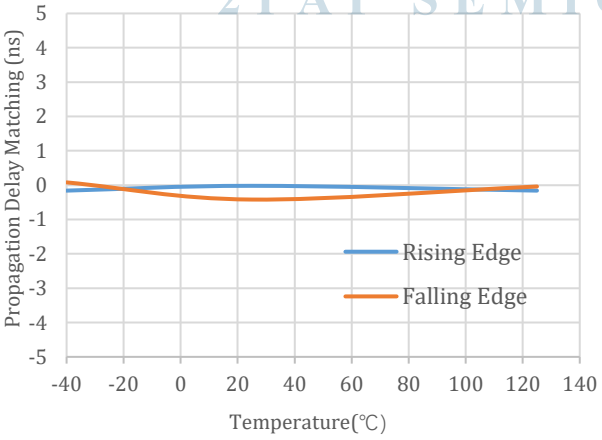


Fig 18.Propagation Delay Matching (t_{DM})

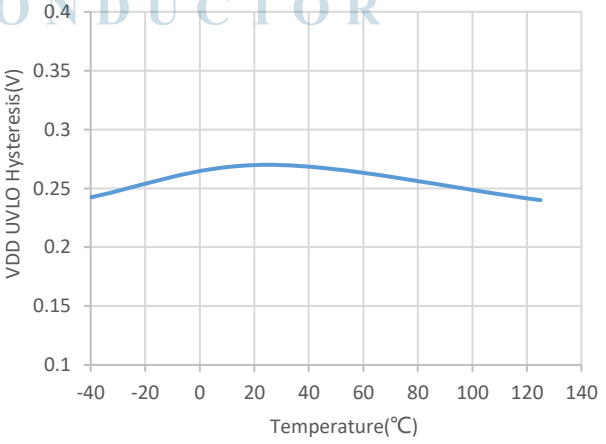


Fig 19.VDD 6-V UVLO Hysteresis

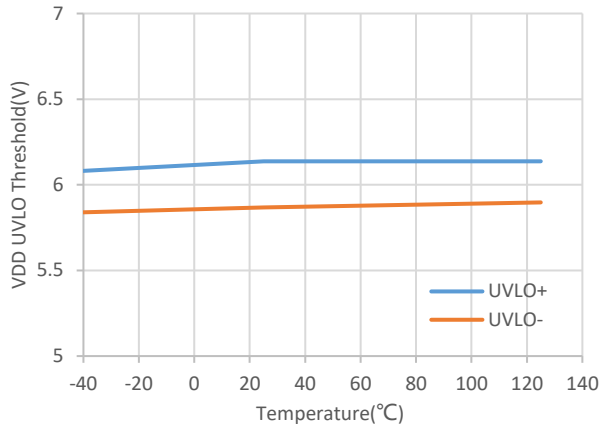


Fig 20.VDD 6-V UVLO Threshold vs. Temperature

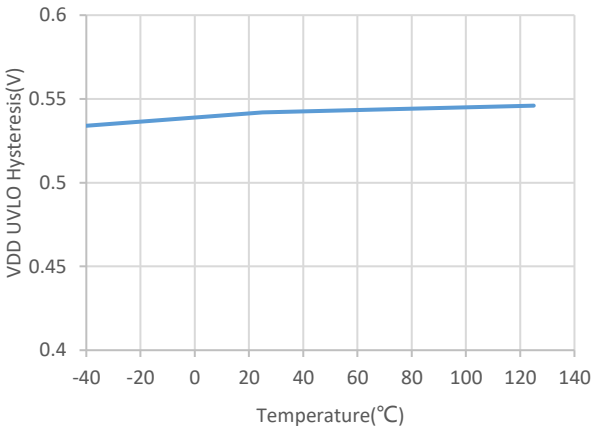


Fig 21.VDD 9-V UVLO Hysteresis

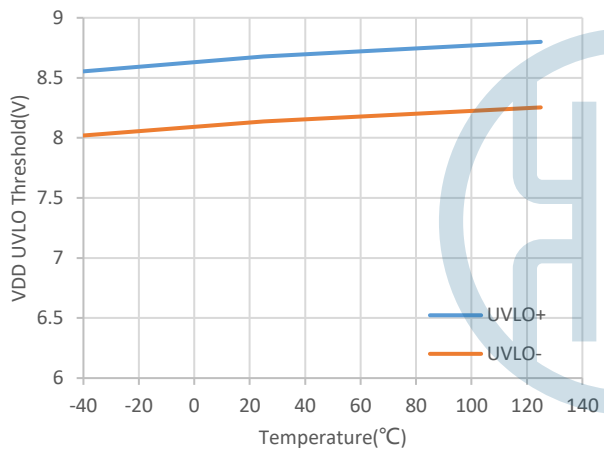


Fig 22.VDD 9-V UVLO Threshold vs. Temperature

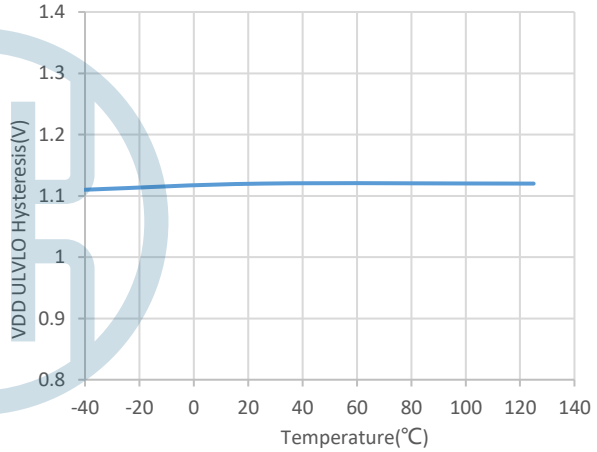


Fig 23.VDD 12-V UVLO Hysteresis

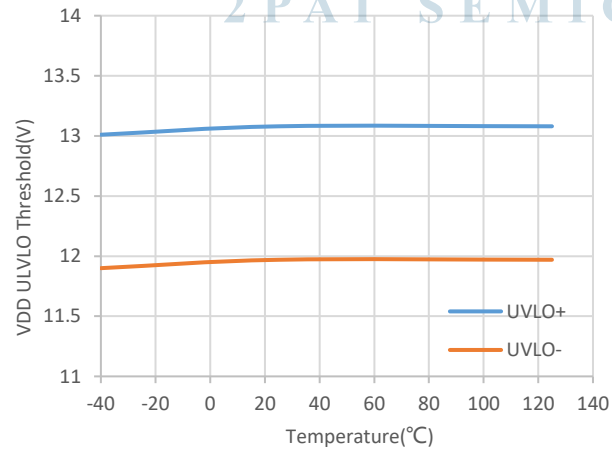


Fig 24.VDD 12-V UVLO Threshold vs. Temperature

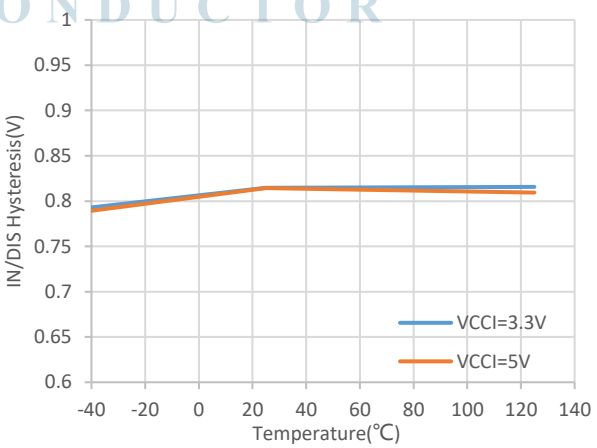


Fig 25.IN/DIS Hysteresis

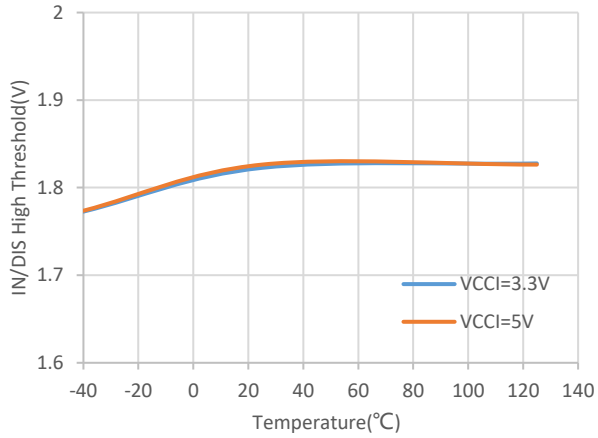


Fig 26.IN/DIS High Threshold (rise)

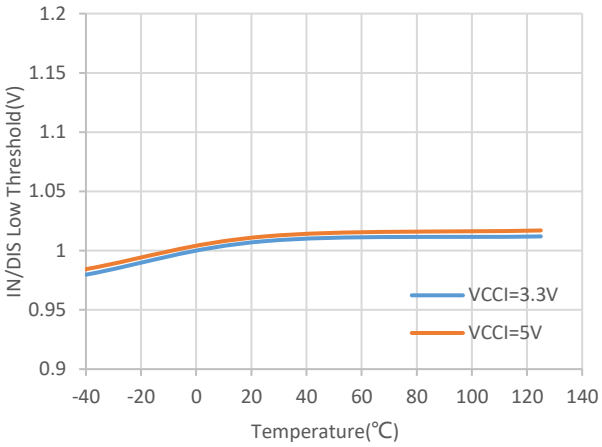


Fig 27.IN/DIS Low Threshold (fall)

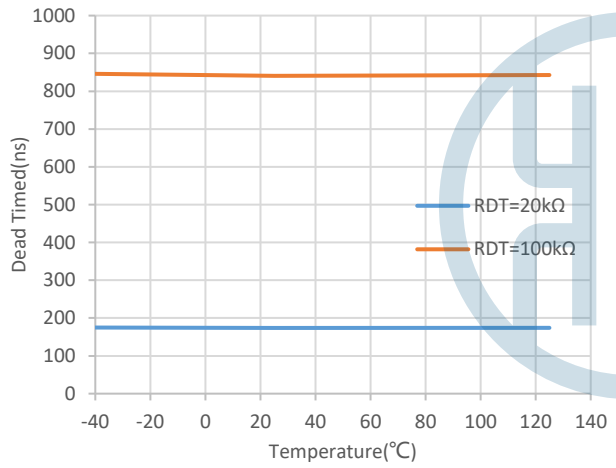


Fig 28.Dead Time

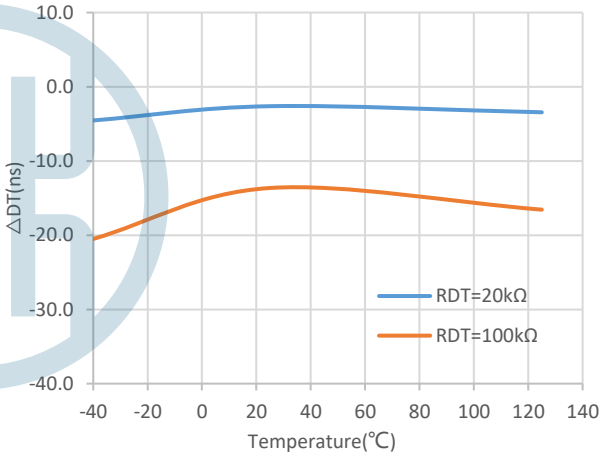


Fig 29.Dead Time Matching

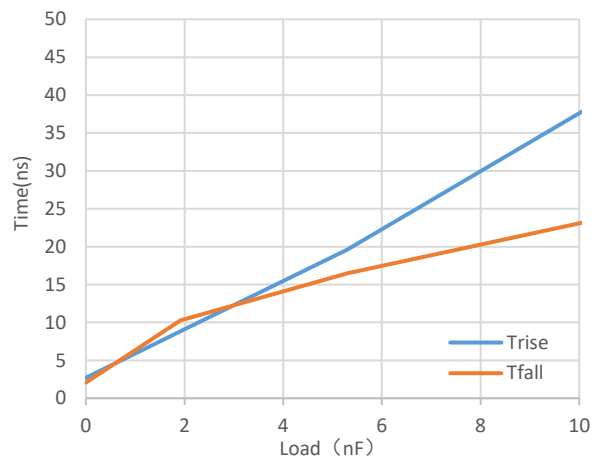


Fig 30.Rising and Falling Time

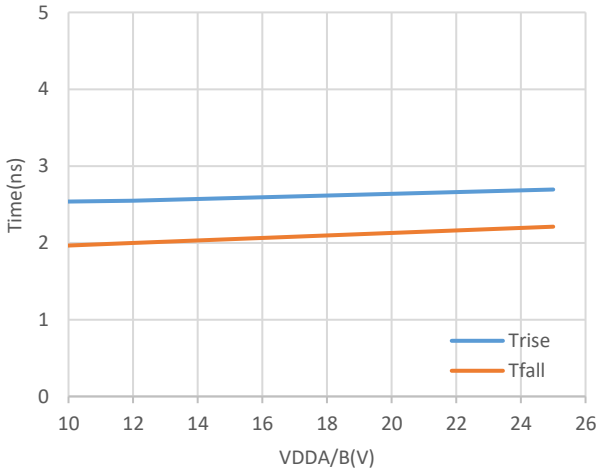


Fig 31.Rise/Fall Time

6. Parameter Measurement Information

6.1.Propagation Delay Matching and Pulse Width Distortion

Fig 32 shows how to calculate pulse width distortion(t_{PWD})and delay matching (t_{DM}) from the propagation delays of channels A and B. It can be measured by ensuring that both inputs are in phase and disabling the dead time function by shorting the DT Pin to VCCI.

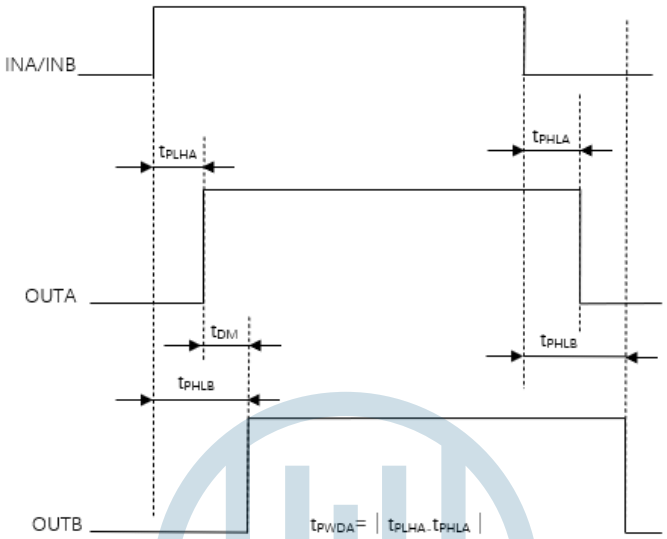


Fig 32.Overlapping inputs, Dead Time Disable

6.2.Rising and Falling Time

Fig 33 shows the criteria for measuring rising (t_r) and falling (t_f) times.

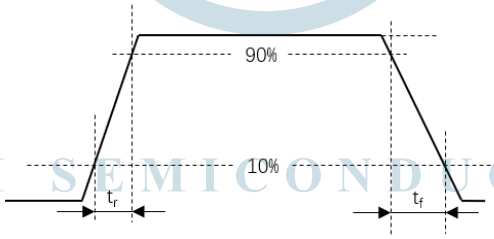


Fig 33.Rising and Falling Time Criteria

6.3.Input and Disable Response Time

Fig 34 shows the response time of the disabled function. It is recommended to bypass using a 1nF low ESR/ESL capacitor close to DIS pin when connecting DIS pin to a micro controller with distance.

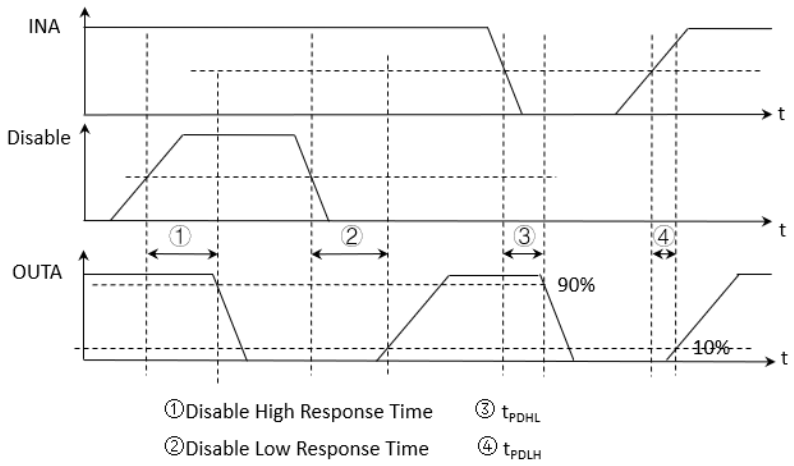


Fig 34.Disable Pin Timing

6.4.Programable Dead Time

Leaving the DT pin open or tying it to GND through an appropriate resistor (R_{DT}) sets a dead-time interval.

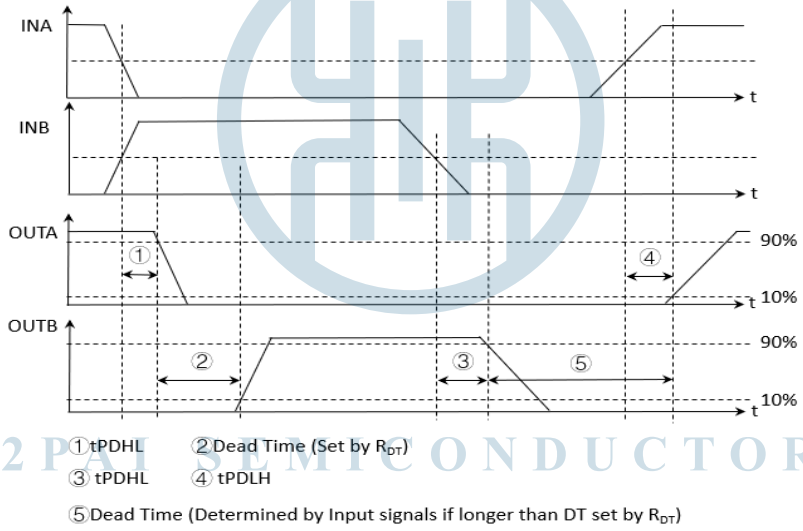


Fig 35.Dead Time Switching Parameters

6.5.Power-up UVLO Delay to OUTPUT

Before the driver is ready to deliver a proper output state, there is a power-up delay from the UVLO rising edge to output and it is defined as $t_{VCCI+\ to\ OUT}$ for VCCI UVLO (typically 82us) and $t_{VDD+\ to\ OUT}$ for VDD UVLO (typically 45us). It is recommended to consider proper margin before launching PWM signal after the driver's VCCI and VDD bias supply is ready. Fig 36 and Fig 37 show the power up UVLO delay timing diagram for VCCI and VDD.

If INA or INB are active before VCCI or VDD have crossed above their respective on thresholds, the output will not update until $t_{VCCI+\ to\ OUT}$ or $t_{VDD+\ to\ OUT}$ after VCCI or VDD crossing its UVLO rising threshold. However, when either VCCI or VDD receive a voltage less than their respective off thresholds, there is <1μs delay, depending on the voltage slew rate on the supply pins, before the outputs are held low. This asymmetric delay is designed to ensure safe operation during VCCI or VDD brownouts.

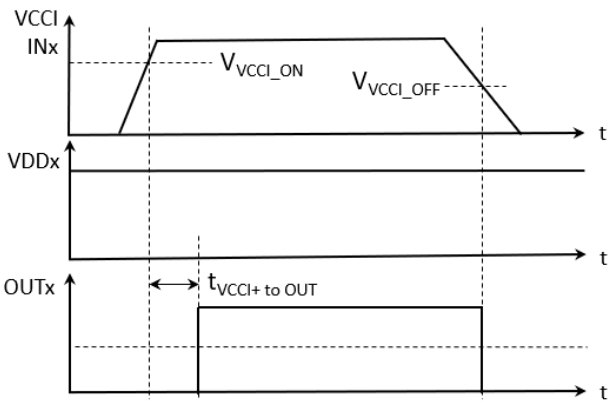


Fig 36. VCCI Power-up UVLO Delay

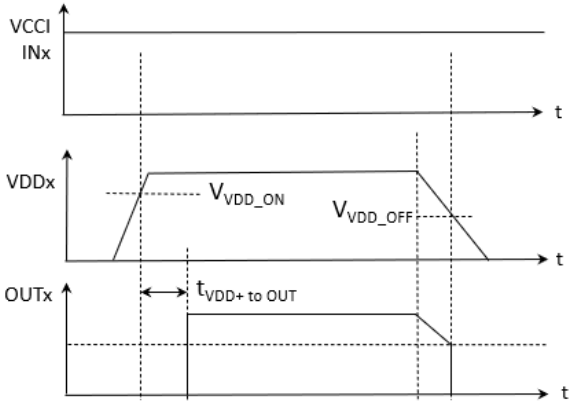


Fig 37. VDDA/B Power-up UVLO Delay

6.6.CMTI Testing

Fig 38 is a simplified diagram of the CMTI test configuration.

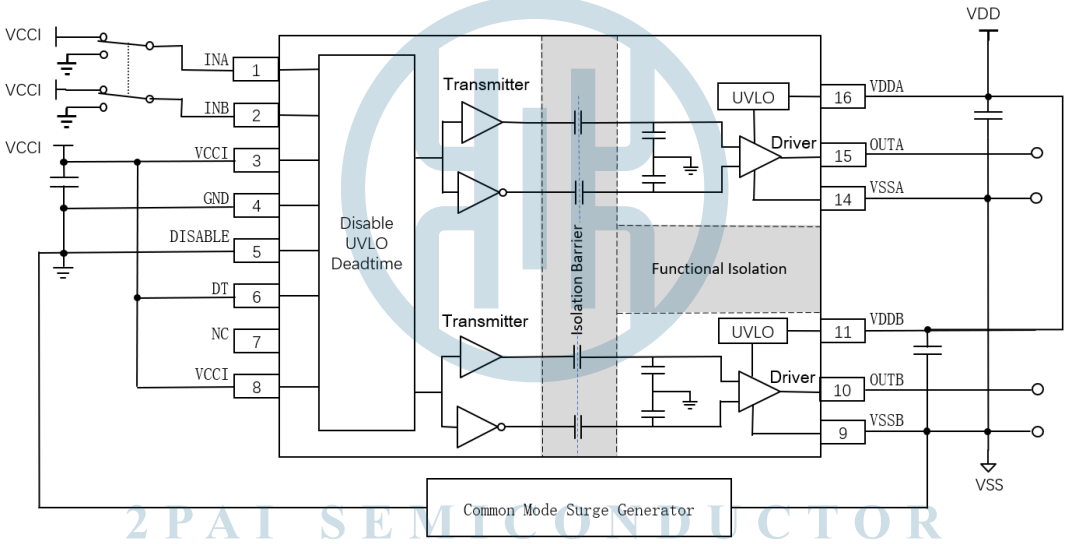


Fig 38.simplified CMTI test setup

7. Detailed Description

7.1.Overview

The Pai8233X/Pai8253X are flexible dual-channel gate drivers. The Pai8233X/Pai8253X have many functions that can make it well integrated with the control circuit and protect the drove devices, such as Pai8233X programmable dead time (DT) control, DISABLE function and undervoltage lockout (UVLO) function. When the input remains open or the input pulse width is insufficient, Pai8233X /Pai8253X keeps its output low.

7.2.Feature Description

7.2.1. VDD, VCCI, and Under Voltage Lock Out (UVLO)

The Pai8233X/Pai8253X have under-voltage protection (UVLO) function in primary and secondary side. When the VDD voltage is lower than V_{VDD_ON} at device start-up or lower than V_{VDD_OFF} after start-up, the VDD UVLO feature holds the effected output low, regardless of the status of the input pins (INA and INB).

Table 14. VCCI UVLO Feature Logic

CONDITION	INPUTS		OUTPUTS	
	INA	INB	OUTA	OUTB
VCCI-GND < V_{VCCI_ON} during device start up	H	L	L	L
	L	H	L	L
	H	H	L	L
	L	L	L	L
VCCI-GND < V_{VCCI_OFF} after device start up	H	L	L	L
	L	H	L	L
	H	H	L	L
	L	L	L	L

Table 15. VDDA/B UVLO Feature Logic

CONDITION	INPUTS		OUTPUTS	
	INA	INB	OUTA	OUTB
VDDA/B -VSSA/B < V_{VDD_ON} during device start up	H	L	L	L
	L	H	L	L
	H	H	L	L
	L	L	L	L
VDDA/B -VSSA/B < V_{VDD_OFF} after device start up	H	L	L	L
	L	H	L	L
	H	H	L	L
	L	L	L	L

7.2.2. Input and Output Logic Table

Table 16. INPUT/OUTPUT Logic Table (Assume VCCI, VDDA, VDDB are powered up.)

INPUTS		DISABLE	OUTPUTS		NOTE
INA	INB		OUTA	OUTB	
L	L	L or Left Open	L	L	If Dead Time function is used, output transitions occur after the dead time expires.
L	H	L or Left Open	L	H	
H	L	L or Left Open	H	L	
H	H	L or Left Open	L	L	DT is left open or programmed with R_{DT}
H	H	L or Left Open	H	H	DT pin pulled to VCCI
Left Open	Left Open	L or Left Open	L	L	-
X[1]	X[1]	H	L	L	-

(1) "X" means L, H or left open.

7.3.Disable Function

Setting the DISABLE pin high shuts down both outputs simultaneously. Grounding (or left open) the DISABLE pin allows the Pai8233X to operate normally. The DISABLE response time is as fast as propagation delay. The DISABLE pin is only functional (and necessary) when VCCI stays above the UVLO threshold. It is recommended to tie this pin to ground if the DISABLE pin is not used to achieve better noise immunity, and it is recommended to bypass using a ≈1nF low ESR/ESL capacitor close to DIS pin when connecting DIS pin to a micro controller with distance.

7.4.Programmable Dead Time (DT) Function

The Pai8233X allows the user to adjust dead time (DT) in the following ways:

7.4.1. Tying the DT Pin to VCC

Outputs completely match inputs, so no dead time is asserted. This allows outputs to overlap.

7.4.2. DT Pin Left Open or Connected to a Programming Resistor between DT and GND Pins

If the Pai8233B/C/E DT pin is left open, the dead time duration (t_{DT}) is set to <15 ns. One can program t_{DT} by placing a resistor, R_{DT}, between the DT pin and GND. The appropriate R_{DT} value can be determined from Equation 1, where R_{DT} is in kΩ and t_{DT} is in ns.

$$t_{DT} \approx 10 \times R_{DT} \quad (1)$$

An input signal's falling edge activates the programmed dead time for the other signal. The output signals' dead time is always set to the longer of either the driver's programmed dead time or the input signal's own dead time. If both inputs are high simultaneously, both outputs will immediately be set low. This feature is used to prevent shoot-through, and it doesn't affect the programmed dead time setting for normal operation. Various driver dead time logic operating conditions are illustrated and explained in Fig 39.

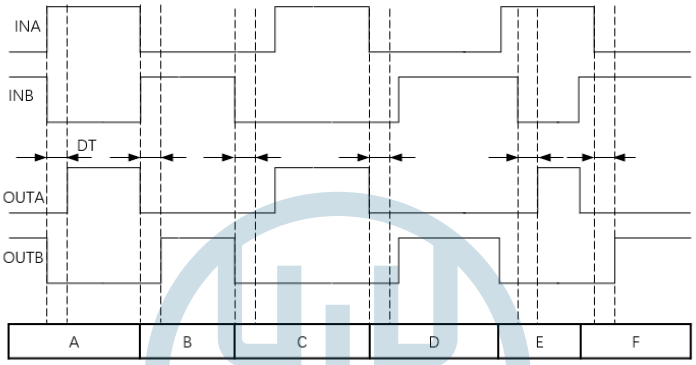


Fig 39.Input and Output Logic Relationship With Input Signals

Table 17.Input and Output Logic Relationship With Input Signals

Condition	INB	INA	Output
A	goes low	goes high	INB sets OUTB low immediately and assigns the programmed dead time to OUTA. OUTA is allowed to go high after the programmed dead time.
B	goes high	goes low	Now INA sets OUTA low immediately and assigns the programmed dead time to OUTB. OUTB is allowed to go high after the programmed dead time.
C	goes low	still low	Condition C: INB goes low, INA is still lowing sets OUTB low immediately and assigns the programmed dead time for Obtain this case, the input signal's own dead time is longer than the programmed dead time. Thus, when INA goes high, it immediately sets OUTA high.
D	still low	goes low	Condition D: INA goes low, INB is still lowing sets OUTA low immediately and assigns the programmed dead time to OUTB.INB's own dead time is longer than the programmed dead time. Thus, when INB goes high, it immediately sets OUTB high.
E	still high	goes high	Condition E: INA goes high, while INB and OUTB are still high. To avoid overshoot, INA immediately pulls OUTB low and keeps OUTA low. After some time INB goes low and assigns the programmed dead time to OUTA. OUTB is already low. After the programmed dead time, OUTA is allowed to go high.
F	goes high	still high	Condition F: INB goes high, while INA and OUTA are still high. To avoid overshoot, INB immediately pulls OUTA low and keeps OUTB low. After some time INA goes low and assigns the programmed dead time to OUTB.OUTA is already low. After the programmed dead time, OUTB is allowed to go high.

8. Typical Application

Note: The information in this section is for reference only, and 2Pai Semi does not guarantee its accuracy or completeness.

8.1.Solution with bootstrap circuits on secondary side and without negative shutdown voltage

Fig 40 shows a reference design in which the Pai8233X/Pai8253X is set as a typical half-bridge drive configuration. In this design, only one independent power supply is required on the secondary side, but there is no negative voltage shutdown feature. This solution is suitable for small and medium power scenarios.

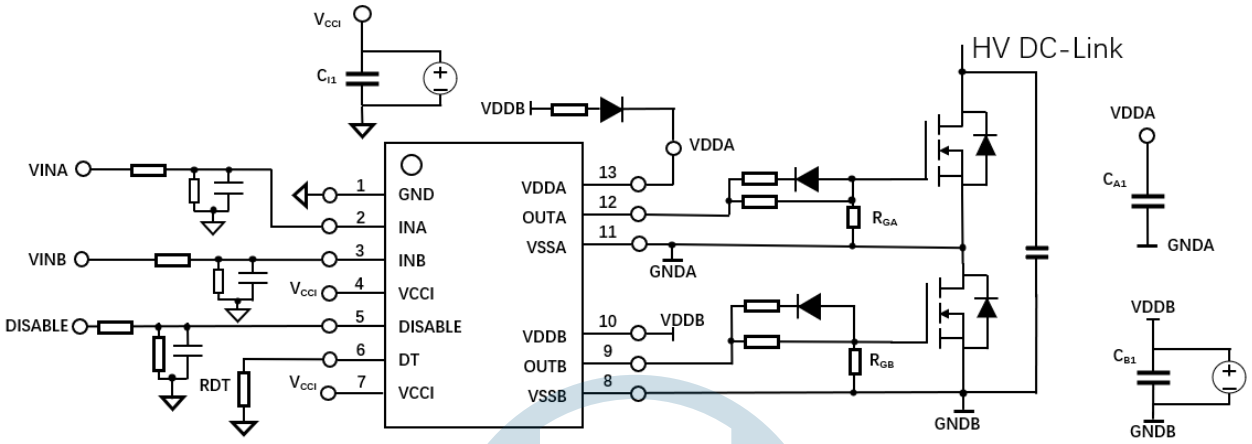


Fig 40.Solution with bootstrap circuits on secondary side and without negative shutdown voltage

8.2.Enhance anti-interference ability solutions with Single power supply and bootstrap circuits for the secondary side

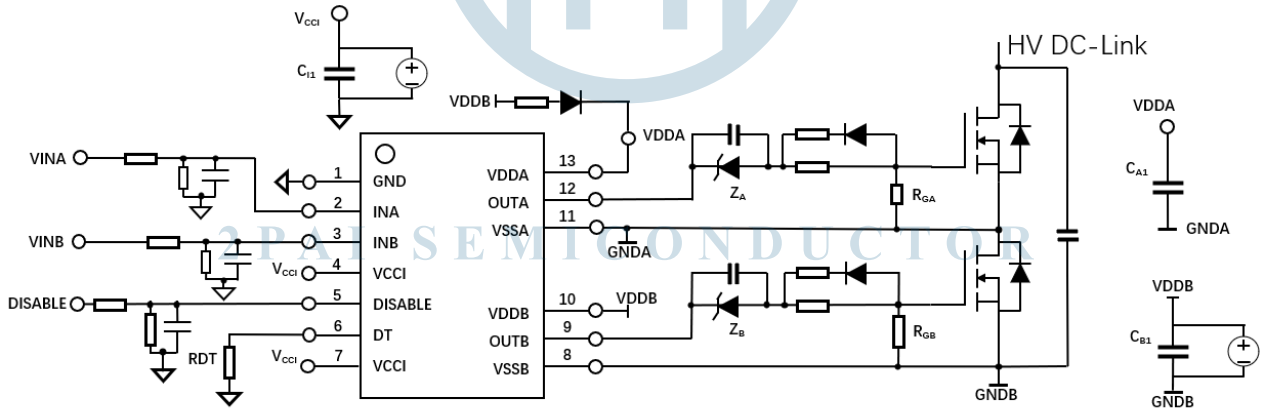


Fig 41.Enhance anti-interference ability solutions with Single power supply and bootstrap circuits for the secondary side

The solution shows in Fig 41 is an improved solution. Only one independent power supply is needed on the secondary side. By adding a Zener diode in the gate drive loop to construct a negative turn-off voltage, the anti-interference ability of the gate is enhanced. This solution has the following limitations:

1) The negative gate bias is not only determined by the Zener diode, but also depends on the duty cycle, which means that when the negative bias duty cycle changes, the negative turn-off voltage will also change.
2) The high side VDDA-VSSA must maintain enough voltage to stay in the recommended power supply range, which means that the low-side switch must be turned on for enough time in each switching cycle.

8.3. Solution with two independent power supplies on the secondary side and a negative turn-off voltage constructed through the Zener diode

When parasitic inductances are introduced by non-ideal PCB layout and long package leads, there would be ringing in the gate-source drive voltage of the power transistor during high di/dt and dv/dt switching. If the ringing is over the threshold voltage, there is the risk of unintended turn-on and even shoot-through. Applying a negative bias on the gate drive is a popular way to keep such ringing below the threshold. Below are a few examples of implementing negative gate drive bias.

Fig 42 shows the example with negative bias turn-off on the channel A driver using a Zener diode on the isolated power supply output stage. The negative bias is set by the Zener diode voltage. If the isolated power supply VA is equal to 25 V, the turn-off voltage will be –5.1 V and turn-on voltage will be 20 V.

The channel-B driver circuit is the same as channel-A, therefore, this configuration needs two power supplies for a half-bridge configuration.

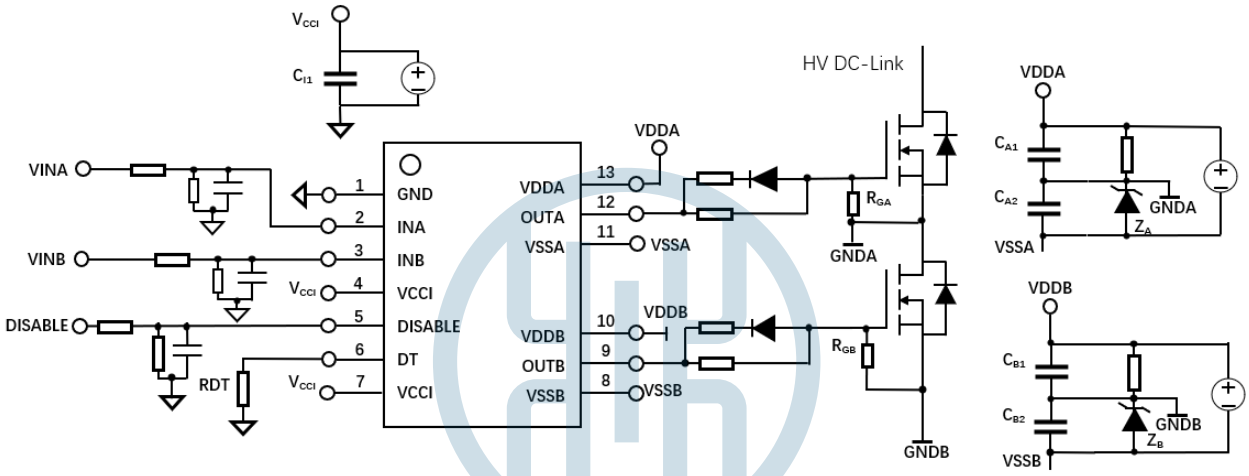


Fig 42. Negative Bias with Zener Diode on Iso-Bias Power Supply Output

8.4.Solution with two independent/four-way power supplies on the secondary side and a negative shutdown voltage

Fig 43 shows a Solution with two independent power supplies on the secondary side and a negative turn-off voltage.

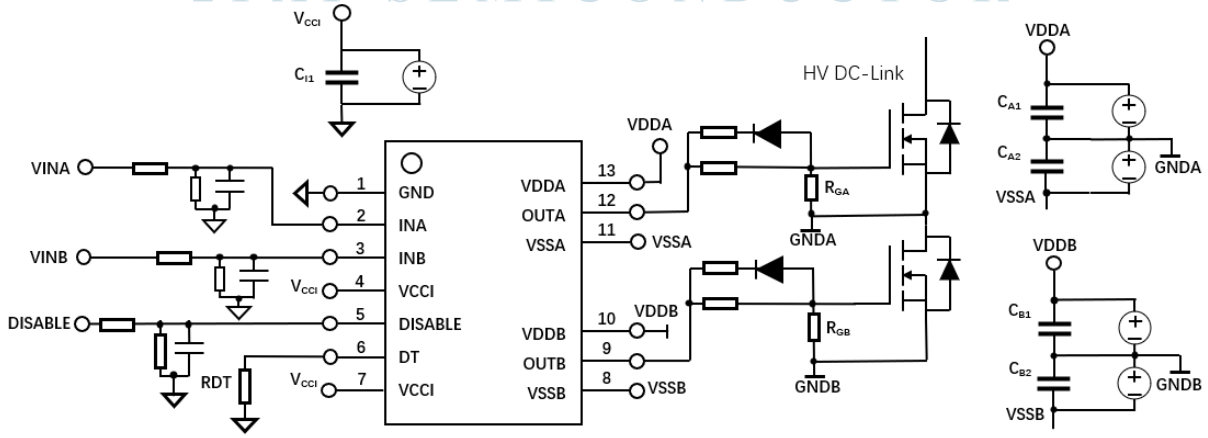
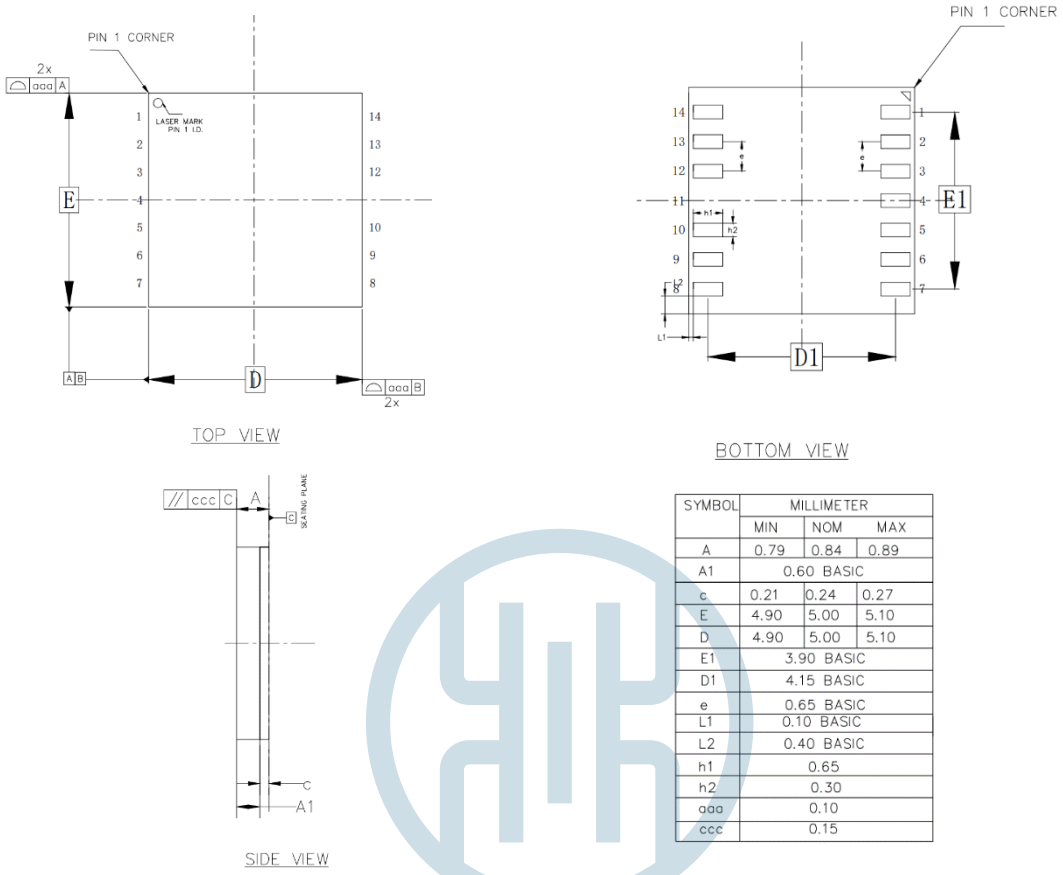


Fig 43. Negative Bias with Two Iso-Bias Power Supplies

9. Outline Dimensions



SYMBOL	MILLIMETER		
	MIN	NOM	MAX
A	0.79	0.84	0.89
A1		0.60 BASIC	
c	0.21	0.24	0.27
E	4.90	5.00	5.10
D	4.90	5.00	5.10
E1		3.90 BASIC	
D1		4.15 BASIC	
e		0.65 BASIC	
L1		0.10 BASIC	
L2		0.40 BASIC	
h1		0.65	
h2		0.30	
aaa		0.10	
ccc		0.15	

Fig 44. LGA13 Outline Package

10. Land Patterns

The Fig 45 illustrates the recommended land pattern details for the Pai8233X/Pai825X in a 13-pin SOIC package. The table lists the values for the dimensions shown in the illustration.

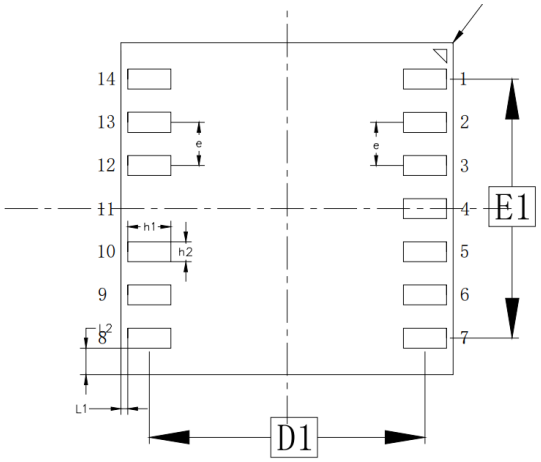


Fig 45.13-Lead LGA13 Land Pattern

Table 18. LGA13 Land Pattern Dimensions

Dimension	Feature	Parameter	Unit
D1	Pad column spacing	4.15	mm
e	Pad row pitch	0.65	mm
h2	Pad width	0.30	mm
h1	Pad length	0.65	mm

Note：
1.This land pattern design is based on IPC -7351
2.All feature sizes shown are at maximum material condition and a card fabrication tolerance of 0.05 mm is assumed.

11.Top Marking

Fig 46.Top Marking

Line 1	PaiXXXXXX=Product name
Line 2	YY = Work Year WW = Work Week ZZ=Manufacturing code from assembly house
Line 3	XXXXX, no special meaning

12.Reel Information

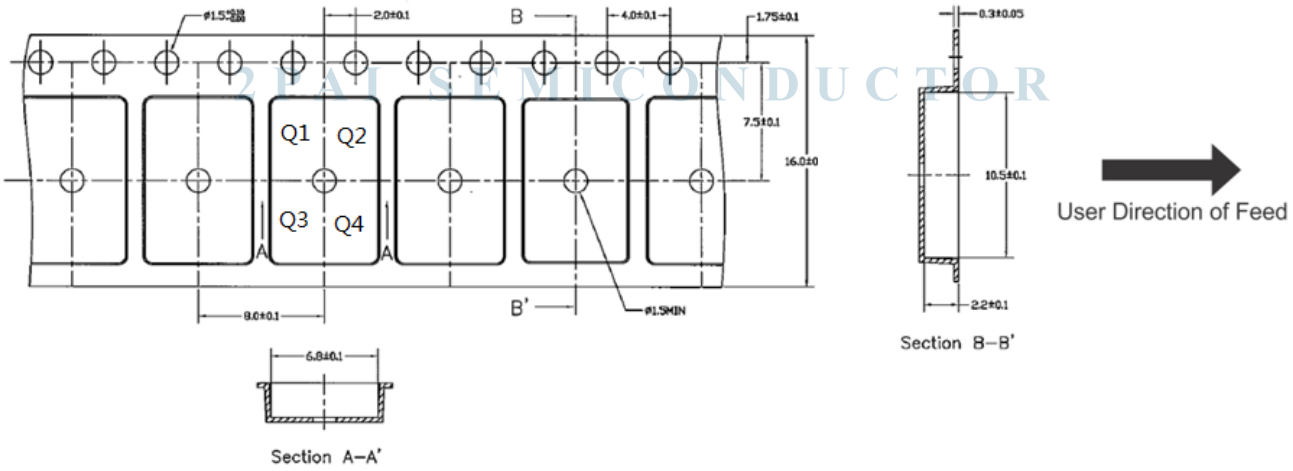


Fig 47. LGA13 Reel Information–dimension unit(mm)

Note：The Pin 1 of the chip is in the quadrant Q1

13. Ordering Guide

Table 21.Ordering Guide

Model Name	Temperature Range	Peak Current	Rec. VDD Supply Min.	Isolation Rating	Package	MSL Peak Temp [1]	Quantity per reel
Pai8233A- LR	−40~125°C	4A/8A	9.2V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8233B- LR	−40~125°C	4A/8A	6.5V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8233C- LR	−40~125°C	4A/8A	9.2V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8233D- LR	−40~125°C	4A/8A	6.5V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8233E- LR	−40~125°C	4A/8A	13.7V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8233F- LR	−40~125°C	4A/8A	13.7V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8253A- LR	−40~125°C	4A/8A	9.2V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8253B- LR	−40~125°C	4A/8A	6.5V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8253C- LR	−40~125°C	4A/8A	9.2V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8253D- LR	−40~125°C	4A/8A	6.5V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8253E- LR	−40~125°C	4A/8A	13.7V	2.5kVrms	LGA13	Level-2-260℃	3000
Pai8253F- LR	−40~125°C	4A/8A	13.7V	2.5kVrms	LGA13	Level-2-260℃	3000

Note：

1. MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

14. Important Notice and Disclaimer

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15. Revision History

Revision	Date	Page	Change Record
Rev.0.1	2022-04-08	All	Initial version
Rev.0.2	2023-03-10	All	Update Pai8253X information